

ALTERNATOR

A.c generators or alternators operate on the same fundamental laws of Faraday's Law of electromagnetic induction. Here armature winding mounted on a stationary element called stator and field winding on a rotating element called Rotor.

The magnetic poles of the rotor are excited from a d.c current source at 125 to 600 volt. When the rotor rotates, the stator conductors being stationary are cut by the magnetic flux, hence they have induced emf produced in them, because the magnetic poles are alternatively north & south, they induce an emf and hence current in armature conductor which first flows in one direction & then in the other. Hence an a.c emf is produced in the stator conductors,

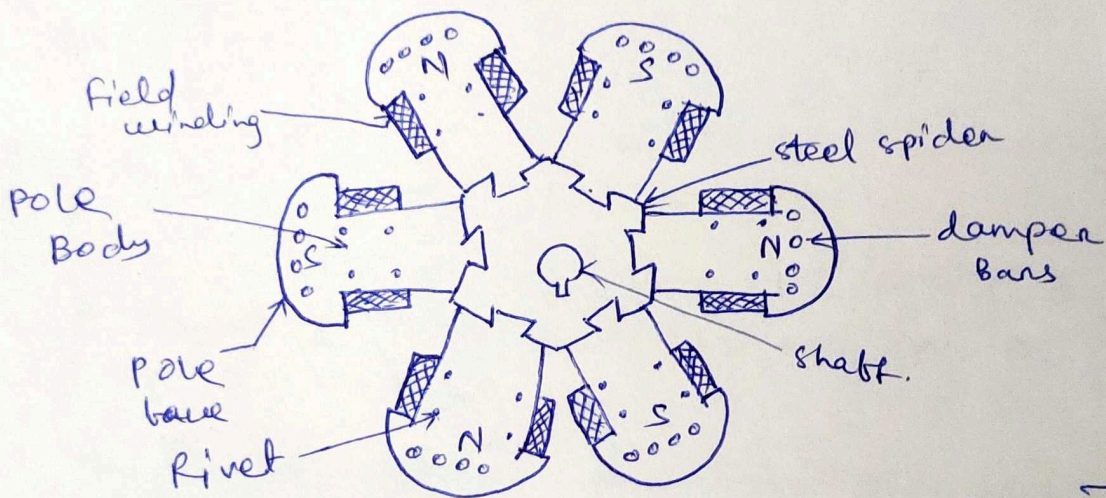
- (i) Whose frequency depends on the no. of north and south poles.
- (ii) Whose direction is given by Fleming's Right hand rule.

Advantages of Stationary Armature:

- (i) The o/p current can be led directly from the fixed terminals of the stator to the load ckt without having to pass it through the brush contacts.

mostly for '2' pole or '4' pole, the pole arcs are surrounded by the field windings placed in slots. To avoid excessive peripheral velocity such rotors have very small dia. Hence, turbogenerators are characterised by small dia, and very long axial length.

② Salient pole type Rotor!



[Six-pole salient-pole Rotor]

→ It is used in low, medium speed driven alternator. It has large no. of projecting salient poles. It is used in hydro-power plants. Such generators are characterised by their large diameter and short axial length.

Damper windings!

These are the copper bars short-circuited at both ends by heavy copper rings.

(2) Easier to insulate stationary ~~alternator~~ armature winding from high A.C voltages.

The sliding contacts i.e. slip rings are transferred to the low voltage, low power d.c field circuit which can therefore easily be insulated.

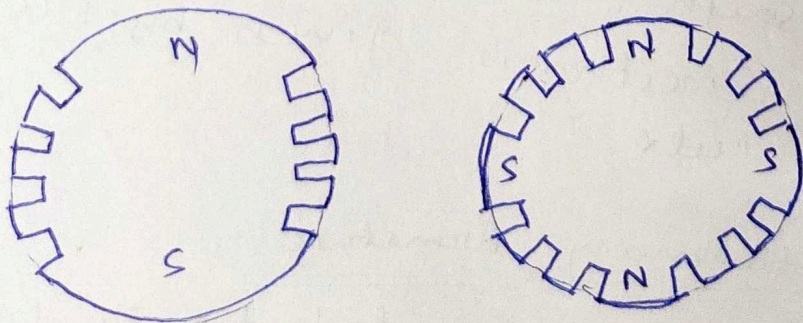
Construction:

Stator: The armature core is supported by the stator frame and is built up of laminations of special magnetic iron or steel alloy. The core is laminated to minimise loss due to eddy current. The laminations are insulated from each other and have spaces between them for allowing the cooling air to pass through. The slots are there for housing the armature conductors.

Rotor: Rotors are of two types.

- (i) Cylindrical type
- (ii) Salient pole type.

(1) Smooth cylindrical type Rotor:



(End view of two pole & four-pole cylindrical rotors)

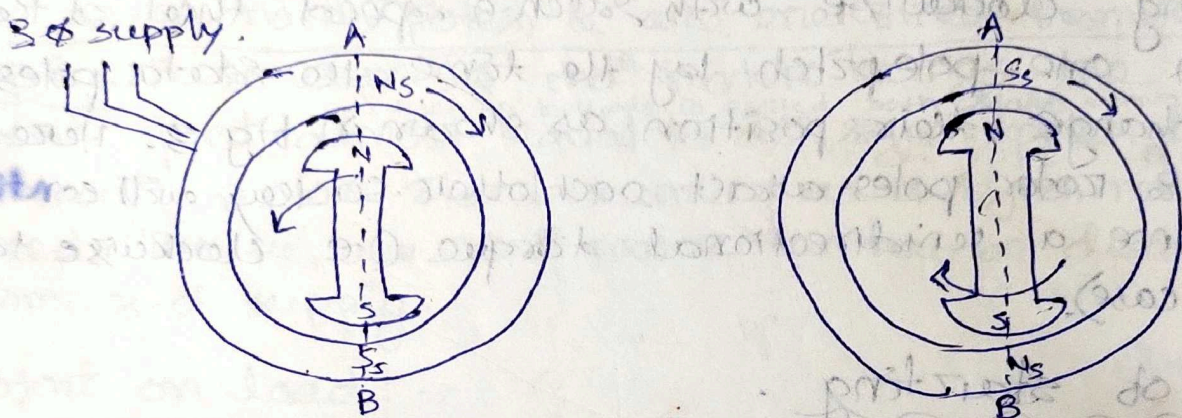
→ It is used for steam turbines driven alternator i.e. turboalternators, which run at high speed. Such rotors are designed

Synch. Motor

A. Mishra

- It runs either at synch. speed or not at all.
- It is not a self started motor. It has to be run up to synch. speed by some means before it can be synchronised to the supply.
- It is capable of being operated under a wide range of p.f. factors.

Principle of operation: →

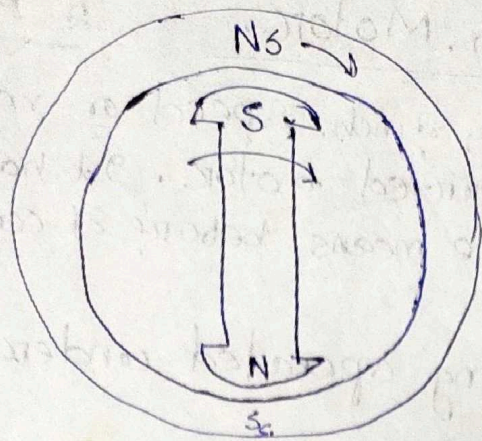


Principle: → (A) (B)

When a 3- ϕ wdg is fed by a 3- ϕ supply, a mag. flux of constant magnitude but rotating at synch. speed is produced.

→ Consider a 2 pole stator, two stator poles N_s & S_s rotate at synch. speed let in clockwise dirⁿ. with the rotor positⁿ as shown. The two similar pole N ' of rotor & N_s ' of stator as well as S and S_s will repel each other with the result that rotor tends to rotate in anticlockwise direction.

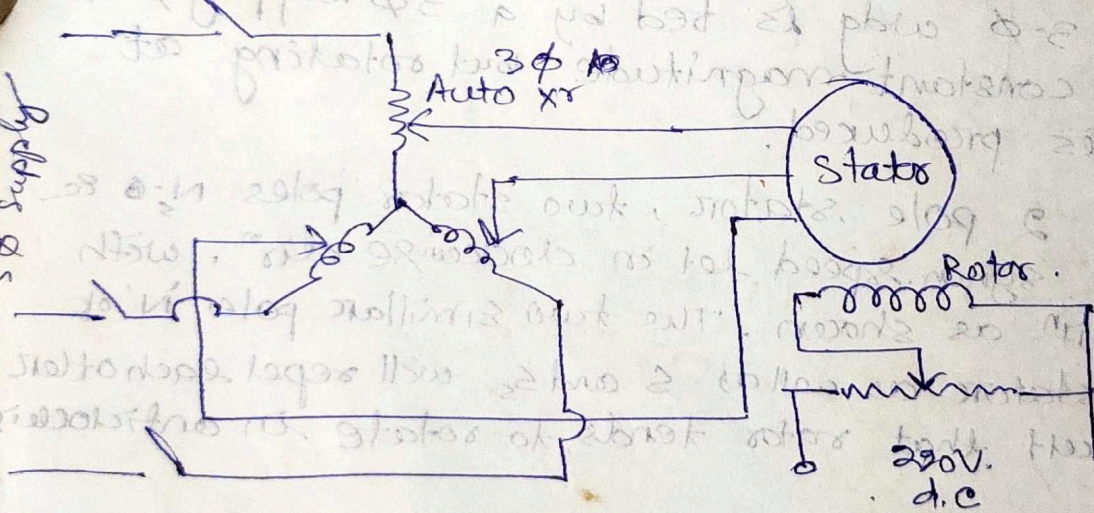
Half a period later stator pole interchange their position (Fig B). At this position N_s and S and S_s & N attract each other with the result that rotor tends to rotate in the clockwise direction. So the rotor is subjected to a torque which is rapidly reversing. Owing to its large inertia the rotor can't instantly response to such quickly reversing torque with the result that it remain stationary.



Suppose that the rotor is not stationary but is rotating clockwise with such a speed that it turns through one pole pitch by the time the stator poles interchange their position as shown in fig 3. Here the stator & rotor poles attract each other. So they will continue to experience a unidirectional torque (i.e. clockwise torque in this case).

Method of starting.

Starting by damper wdg (Induction motor starting)



→ main field wdg short-circuited.
 → reduced volt. with the help of auto Xr is applied across the stator terminals. The motor starts up like an induction motor.

→ when it reaches a steady speed, a weak d.c. excitatⁿ is applied by removing the short-circuit on the main field wdg. If excitatⁿ is sufficient, then the m/c will be put into synchronism.

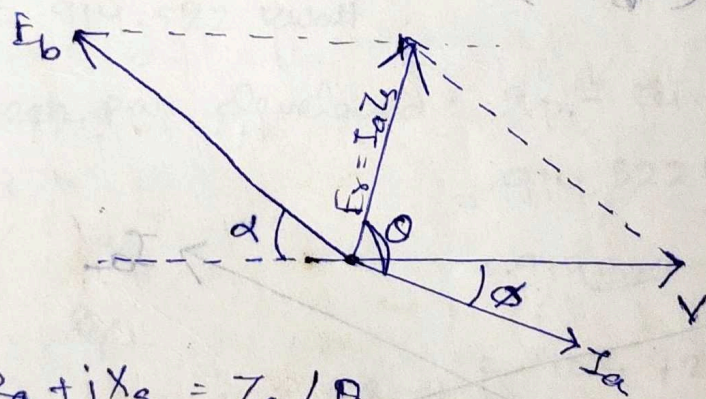
→ Full supplied volt. is applied across stator terminals by cutting out the auto Xr

→ The motor may be operated at any desired p.f by changing the d.c. excitation.

Auxiliary motor starting: →

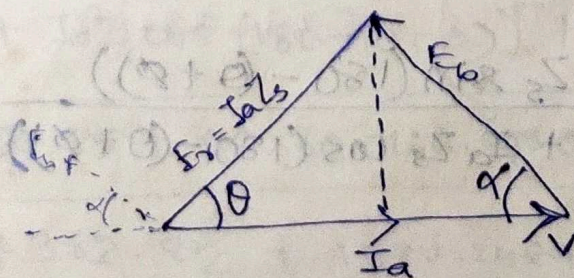
The purpose of the auxiliary motor is to bring the synch. motor speed near to its synch. speed. The aux. motor may be an indⁿ motor or a d.c. motor. when 3- ϕ indⁿ motor is used as an aux. motor, then it is mechanically coupled with synch. motor. Both the motors have same no. of poles & are energized from the same 3 ϕ supply. These aux. 3 ϕ indⁿ motor brings the main motor speed almost equal to its synch. speed. As a result the main motor starts running as a synch. motor at synch. speed. The aux. indⁿ motor can now be disconnected from 3- ϕ supply.

Motor on load: →



$$Z_s = R_a + jX_s = Z_s \angle \theta$$

① unity p.f: →



$$E_b = \sqrt{(V - I_a Z_s \cos \theta)^2 + (I_a Z_s \sin \theta)^2}$$

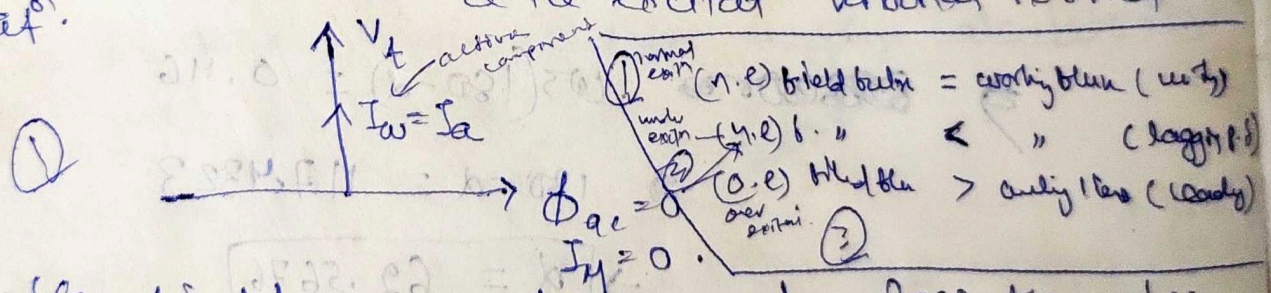
load angle $\tan \alpha = \frac{I_a Z_s \sin \theta}{V - I_a Z_s \cos \theta}$

Power factor control of synch. motor →

An ac ele. magnetiz. device must draw a magnetizing current from the ac source in order to establish the working flux. This magnetizing current lags the applied volt. by almost 90° . A synchronous

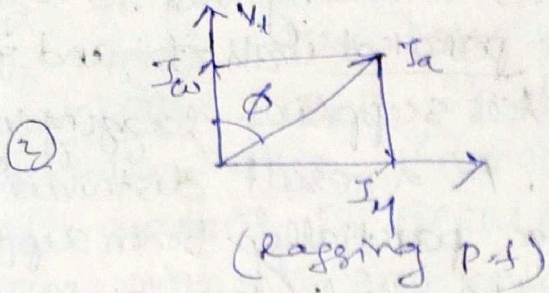
A synch. motor is a doubly excited m/c. Its arm. wdg is excited from an ac source & the field wdg from a dc source. The resultant airgap flux is established by the cooperation of both ac & arm. wdg. & dc in the field wdg, i.e. $\phi_{\text{resultant}} = \phi_{ac} + \phi_{dc}$

i) If the field current is sufficient enough to setup the airgap flux, then magnetizing current required from the ac source is zero and therefore the motor operates at unity p.f. This field current is known as normal field current & the excited known as normal excitation.

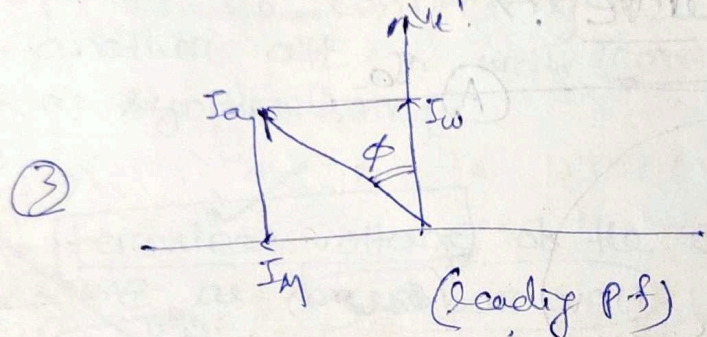


ii) If the field current is made less than the normal excitation, the motor is underexcited. Then the deficiency in the flux must be made up by the arm. wdg. In order to do this, the arm. wdg draws a magnetizing current from the source, and

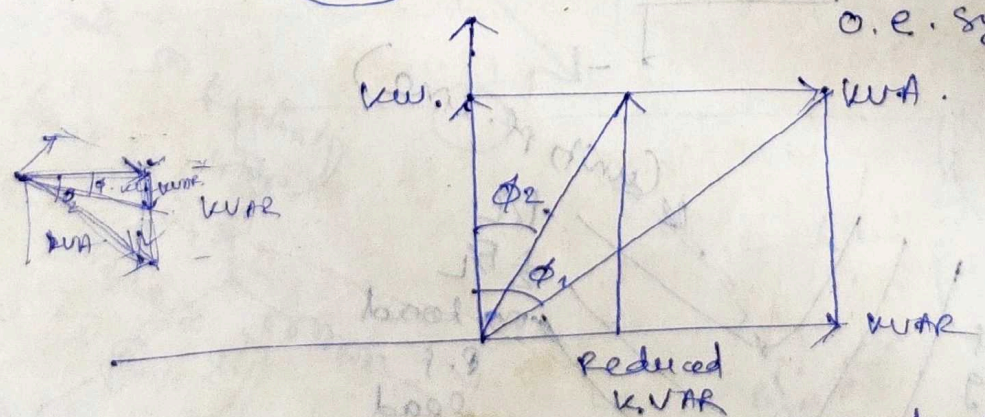
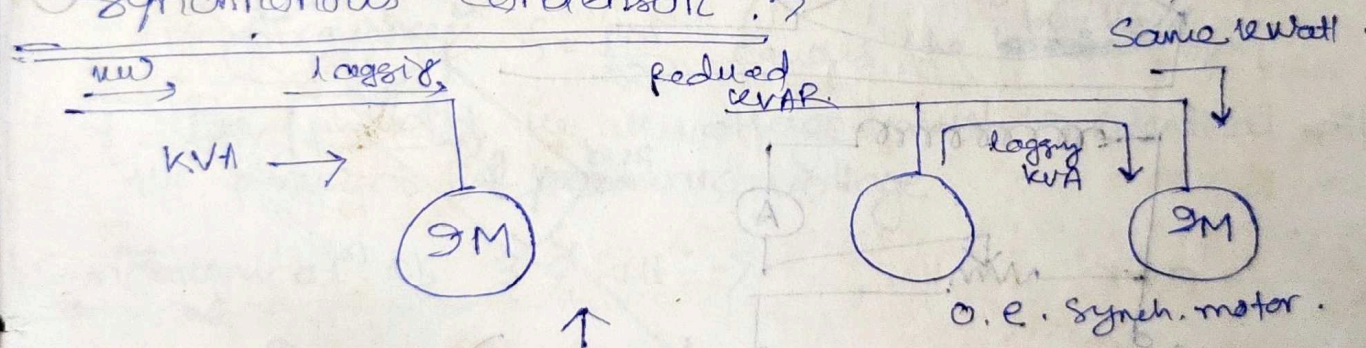
as a result of this, the motor operates at a lagging power factor.



in case the field current is more than its ^{normal} excitation, i.e. motor is overexcited, then the excess flux must be neutralised by the armature wdg. The arm. can do so, only if it draws a demagnetising component of current from the ac source. As magnetising component is lagging behind V by 90° , the demagnetising component must lead V by 90° as a result, the motor operates at a leading pf.



Synchronous Condenser : →

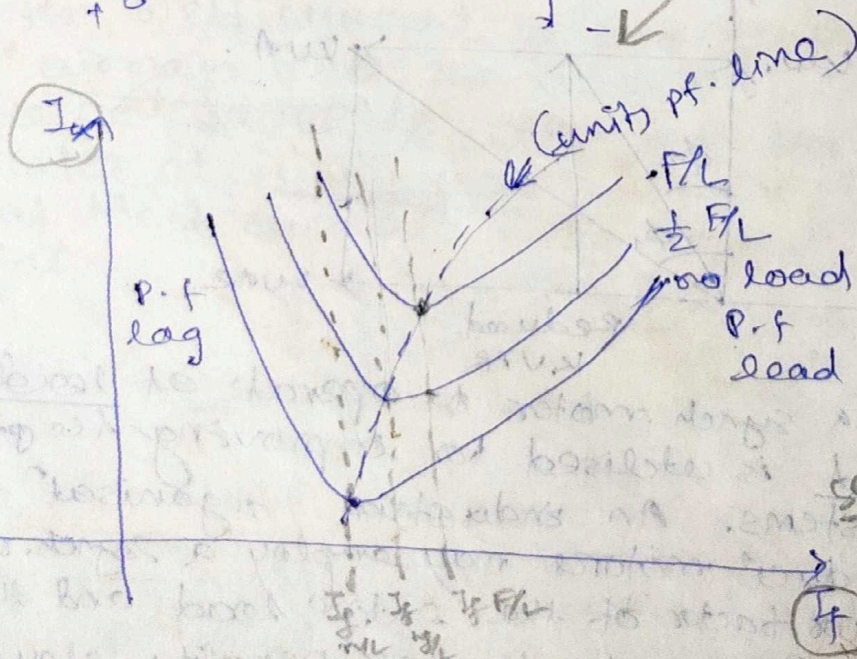
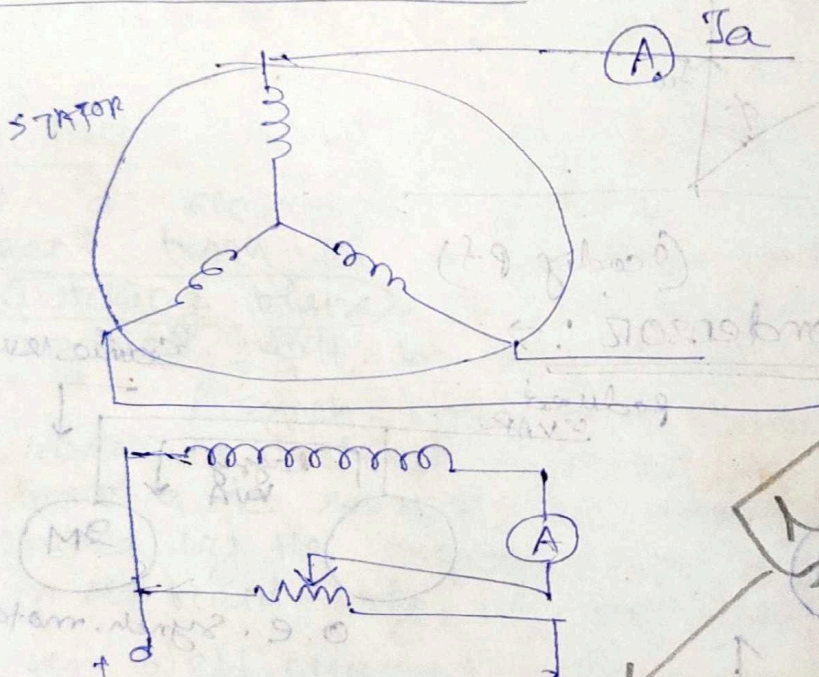


The ability of a synch. motor to operate at leading pf when over excited is utilised for improving the pf. of many supply systems. An industrial organisation having a large no. of inductⁿ motors may employ a synch. motor to improve the power factor of the entire load and thus save the organisation from the low pf penalty clause.

An indⁿ motor must take lagging kVAR from the supply to system to set up its working flux. When a synch. motor is connected in parallel with it, and its this motor is over excited, this supplies lagging kVAR to the indⁿ motor locally. As a result distributed lines are relieved completely or partially from supplying the lagging kVAR needed by the induction motor. As a result P.f. improves. β

So when an over excited unloaded synch. motor is used to improve the P.f. of a system it is known as synch. condenser.

Constructⁿ of V-curve \rightarrow



Load, I_f , P.f., I_a
 ME
 Conductor

(with P.f. of I_a min)
 I_a , I_f lead.
 I_f lag P.f.

same I_f
 FL or $1/2 FL$
 check (lag \rightarrow lead)
 $1/2 FL$ vs FL am
 lag P.f.

It is the curve bet field current I_f to arm. current I_a . In order to draw these curves experimentally the motor is run from the constant volt. & cont. freq. bus bars. pow. sup to motor is kept constant at a definite value. Connection is made like the net diagram. Field current is increased in small steps and corresponding arm. currents are noted. when plotted I_a vs. I_f , we get a curve like 'V' that's why known as 'V' curve.

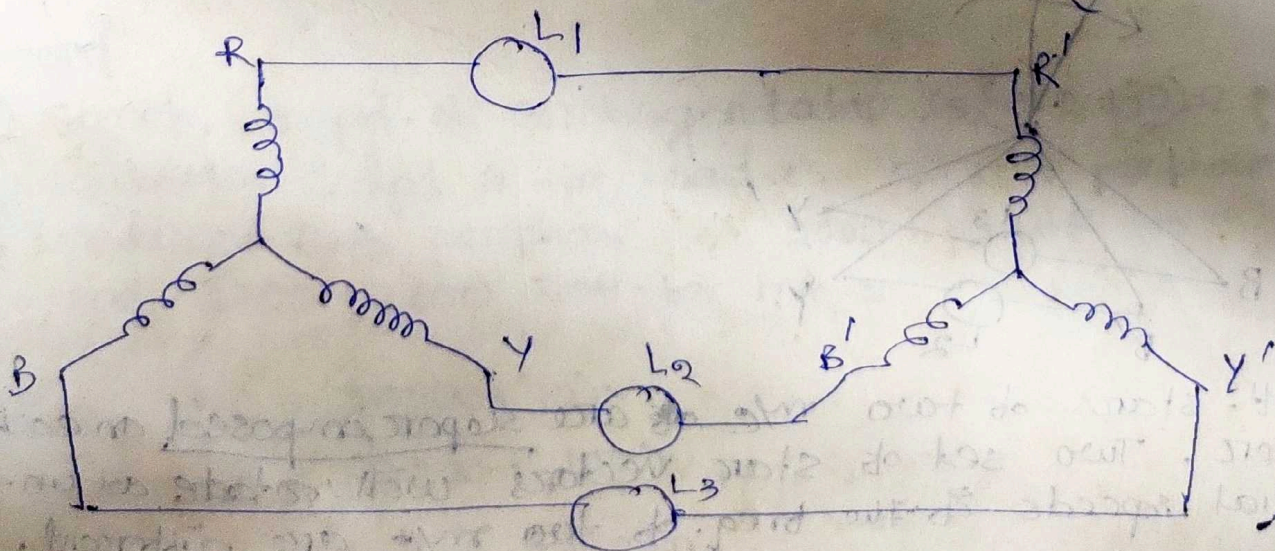
Parallel Operation of alternators

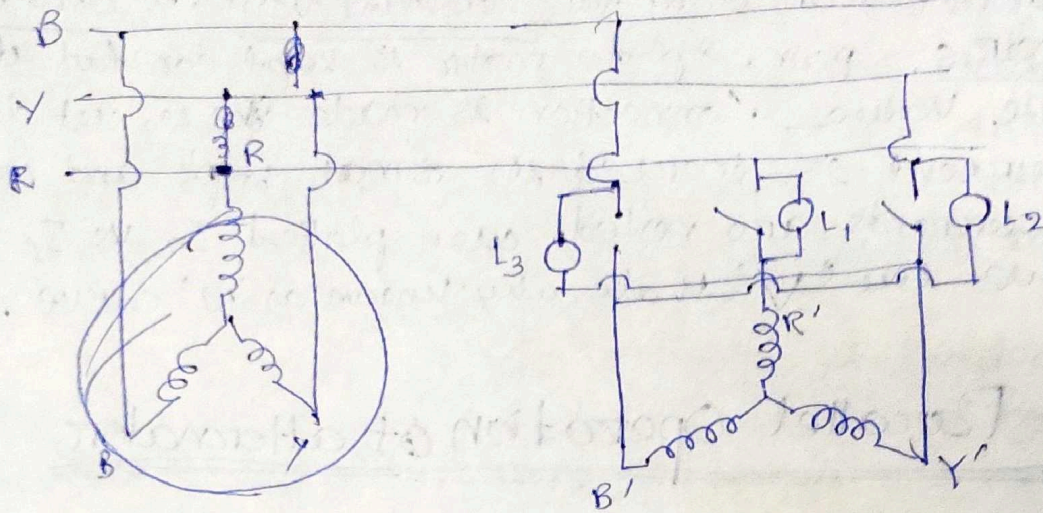
The operatⁿ of connecting an alt. in parallel with another alt. or with common bus bar is known as synchronising.

Condⁿ: →

- (i) The terminal voltage of the incoming alt. must be same as bus bar volt.
- (ii) The speed of the incoming m/c must be such that its frequency $f = \frac{PN}{120}$ equals the bus bar freq.
- (iii) The phase of the alt. voltage must be identical with the phase of the bus bar voltage.

Synchronisatⁿ of 3- ϕ alt.: →

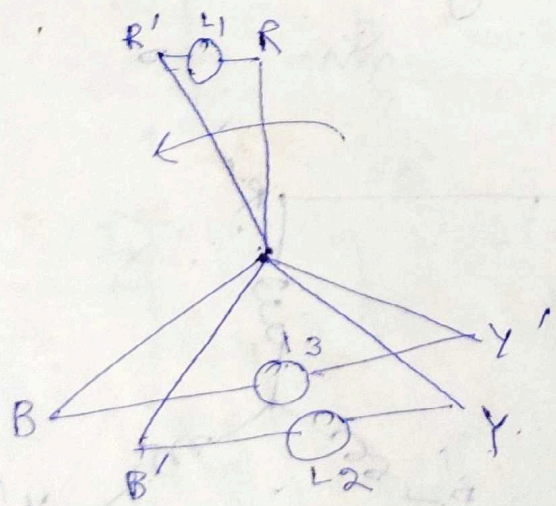




condition (1) is indicated by voltmeter. In 3 ϕ a.c. to synchronise 1 ϕ only, the other 2 phases will then be synchronised automatically.

The incoming a.c. R-Y-B should be properly identified.

→ To test frequency, 3 lamps are used, but they are deliberately connected asymmetrically as shown in the figure. Lamp L_1 is connected betⁿ R & R'. L_2 is connected betⁿ Y & B'. L_3 is connected betⁿ B & Y'.



volt. stars of two m/c are superimposed on each other. Two set of star vectors will rotate at unequal speeds \because the freq. of two m/c are different.

\because the incoming a.c. is running faster, then volt. star R'Y'B' will appear to rotate anticlockwise w.r.t. brush bar volt. star RYB, at a speed corresponding to the

difference betⁿ their frequencies, ~~with~~

As volt. across L_1 is RR' which is increasing from zero, that across L_2 is YB' which is decreasing from its max^m and that across L_3 is BY' which is increasing from zero.

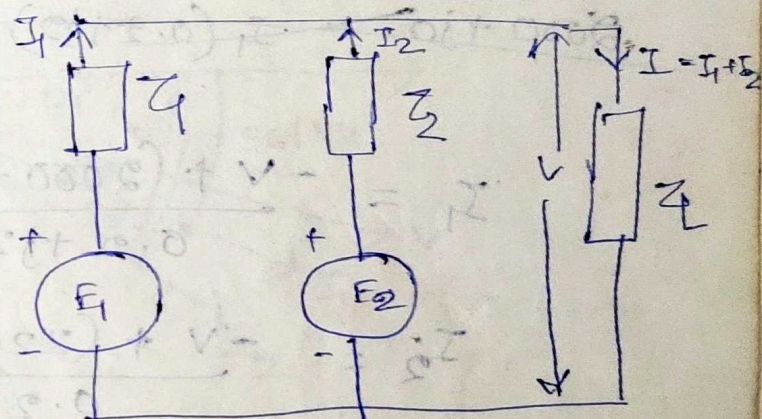
→ Hence the lamp will light up one after the other. When both the m/c breq. are same, uncrossed lamp L_1 will become dark and crossed lamp L_2 & L_3 will glow at same intensity.

→ Synchronizing is done at the moment, the uncrossed lamp L_1 is in the middle of the dark period. When L_1 is dark L_2 & L_3 are equally bright. This method is known as two bright & one dark method.

$$E_1 - I_1 Z_1 = V \quad \text{--- (1)}$$

$$E_2 - I_2 Z_2 = V \quad \text{--- (2)}$$

$$V = I Z_L = (I_1 + I_2) Z_L \quad \text{--- (3)}$$



Two 3- ϕ , synch. mechanically coupled gen^s operated in parallel on the same load. Determine the kWatt ϕ & power factors of each m/c under the following condⁿ

- (i) synch. impd. of each generator is $(0.2 + j2)\Omega$ per phase
- (ii) equivalent impd of the load is $3 + j4\Omega$ per phase.
- (iii) induced emf per phase is $2000 + j0$ volt for m/c-1 and $2200 + j100$ volt for m/c-2.

$$E_1 = (2000 + j0) \text{ V.}$$

$$E_2 = (2200 + j100) \text{ V.}$$

$$Z_1 = Z_2 = (0.2 + j2)\Omega$$

$$Z_L = (3 + j4)\Omega$$

Inductⁿ motor.

It converts ele. pow. to mechanical pow. by
indⁿ motor

Advantage →

- It has very simple & almost unbreakable constructⁿ.
- Its cost is very low.
- It requires less maintenance.
- It has sufficiently high eff.

disadvantage →

- Its speed decreases with increase in load
- Its starting torque is inferior to that of dc shunt motor

Construction →

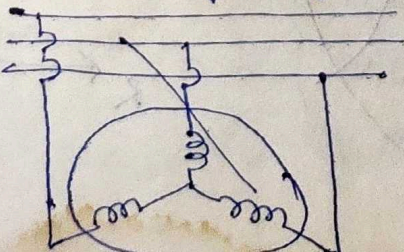
① STATOR → The stator is wound for a definite no. of poles, when supplied with 3 ϕ current it produces a magnetic flux which is of constant magnitude but rotates at synchronous speed $N_s = \frac{120f}{P}$.

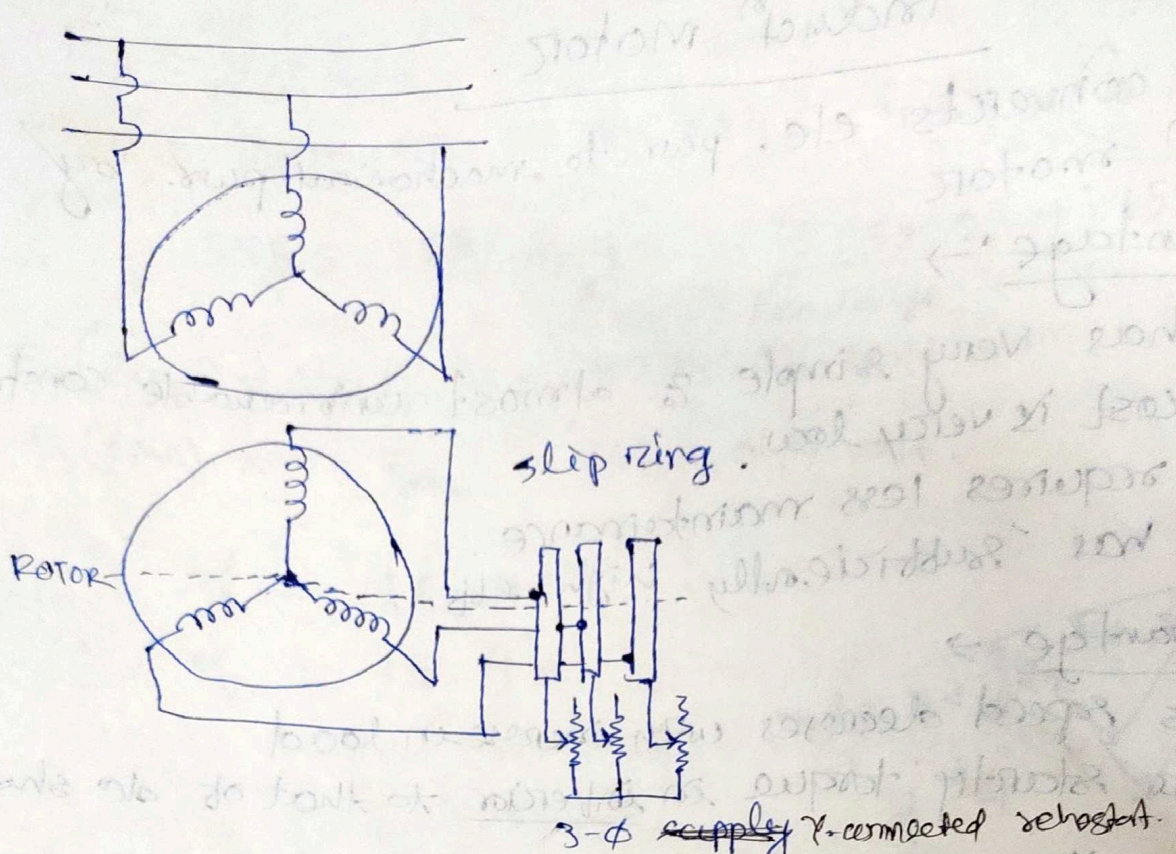
② ROTOR → ① Squirrel cage type rotor →

Inductⁿ motor using squirrel cage type rotor is known as squirrel cage induction motor (SCIM). The rotor consists of a cylindrical laminated core with parallel slots for carrying the rotor conductors which are not wires but consists of heavy bars of copper, aluminium or alloys. The rotor bars are electrically welded or bolted to heavy and stout short-circuiting end rings.

② Slip Ring Rotor →

The inductⁿ motor using this type of rotor is known as slip ring indⁿ motor (SRIM).

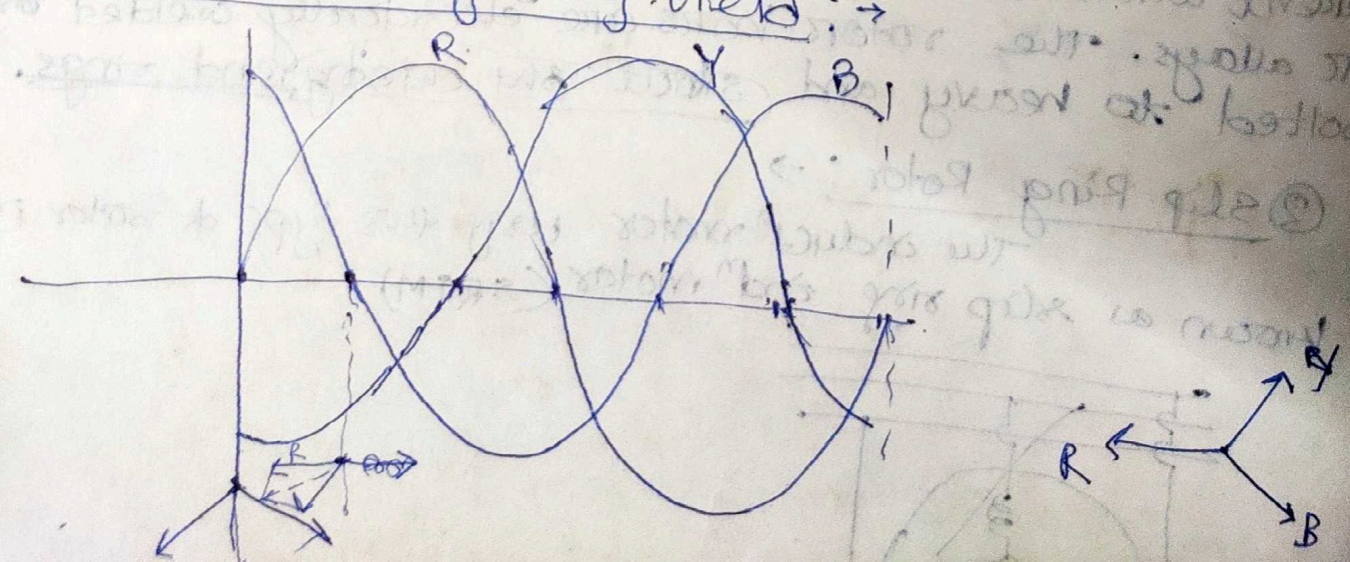




The 3- ϕ of the rotor are starred externally, the other 3 wdg terminals are brought out to connect to 3 insulated slip rings mounted on shaft with brushes resting on them. These 3 brushes are starred externally connected to a 3 ϕ Y-connected resistor. This makes possible the introduction of additional resistance in the rotor cut during starting period by increasing the starting torque of the motor.

Under normal running condⁿ, all the slip rings connected together. So it acts like a squirrel cage type rotor. (shorted)

Product of rotating mag. field \rightarrow



(ii) This resulting flux rotates
by $N_s = \frac{120f}{P}$.

Synch. speed
 $\omega = \frac{4\pi f}{P}$

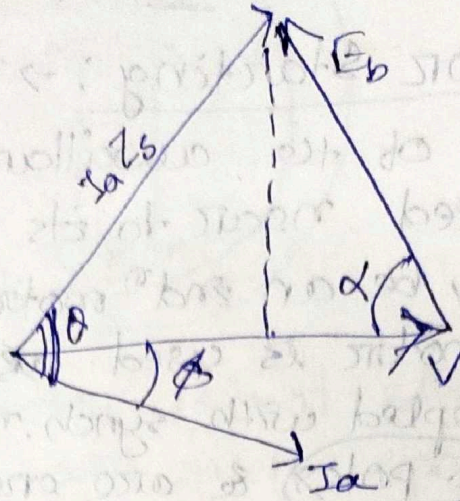
Principle \rightarrow

\rightarrow when 3 ϕ supply is given to the stator it
produce a rotating mag. flux which rotates at sync.
Speed $N_s = \frac{120f}{P}$.

\rightarrow this rotating flux cuts the rotor conductors, so
an emf is induced in it according to Faraday's
law of electromag. Inductⁿ $\mathcal{E} = -N \frac{d\phi}{dt}$

\rightarrow As the rotor is shortcircuited, this emf produces
a current in the rotor conductors. This current
interacts with the flux to produce a torque
(Fleming's left hand rule). As a result rotor rotates.

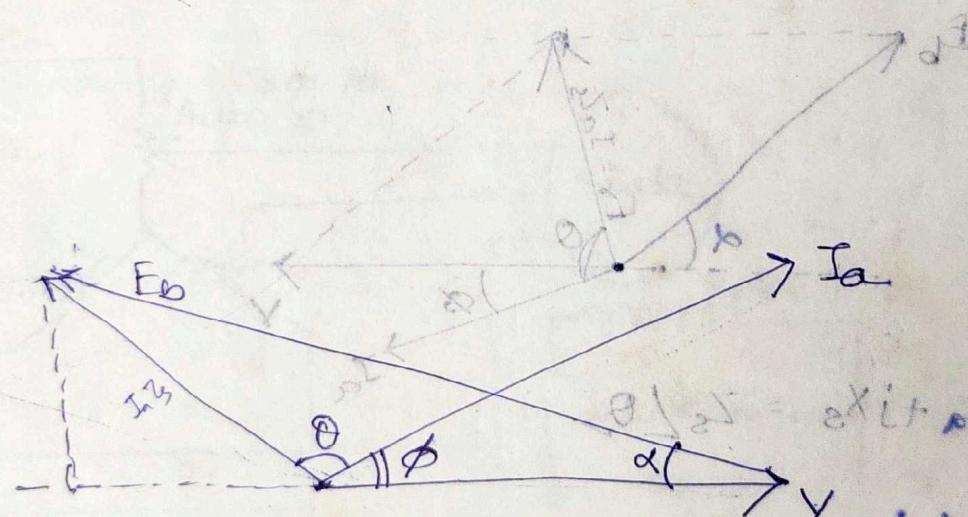
Lagging p.f.



$$E_b = \sqrt{(V - I_a Z_s \cos(\theta - \phi))^2 + (I_a Z_s \sin(\theta - \phi))^2}$$

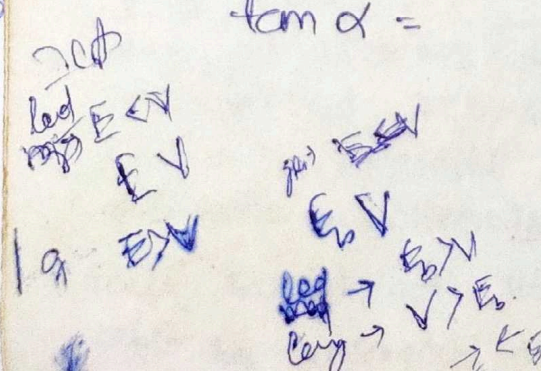
(load angle α is \angle between E_b and V)
 $\tan \alpha = \frac{I_a Z_s \sin(\theta - \phi)}{V - I_a Z_s \cos(\theta - \phi)}$

Leading p.f. ($E_b > V$)



$$E_b = \sqrt{\{V + I_a Z_s \cos(180 - (\theta + \phi))\}^2 + \{I_a Z_s \sin(180 - (\theta + \phi))\}^2}$$

$$\tan \alpha = \frac{I_a Z_s \sin(180 - (\theta + \phi))}{V + I_a Z_s \cos(180 - (\theta + \phi))}$$



Here ele. pow is converted into mechanical pow not by conduction, but by induction like a X^0 .
 so sometime known as rotating X^0 .

Slip: In practice rotor never succeeds in catching up with stator field. If it did so, there would be no relative speed betⁿ the two, hence no rotor emf, no rotor curr, & so no torque to maintain rotation.

The diff. betⁿ synch. speed (N_s) & Actual speed (N_r) of the rotor is known as slip. Express as percentage of synch. speed. (Actually the term 'slip' is descriptive of the way in which the rotor slips back from synchronism.

$$S = \frac{N_s - N_r}{N_s}$$

$$\% \text{ slip} = \frac{N_s - N_r}{N_s} \times 100.$$

$$S = 1 - \frac{N_r}{N_s}$$

$$\Rightarrow \frac{N_r}{N_s} = 1 - S$$

$$\Rightarrow \boxed{N_r = (1 - S) N_s}$$

$$\text{slip speed} = N_s - N_r$$

Frequency of rotor current: \rightarrow

let f = supply frequency

$$N_s = \frac{120f}{P} \quad \text{--- (1)}$$

let f' \rightarrow rotor current freq.
 = rotor emf frequency.

$$N_s - N_r = \frac{120f'}{P} \quad \text{--- (2)}$$

$$\Rightarrow \frac{N_s - N_r}{N_s} = \frac{f'}{f}$$

$$\Rightarrow \frac{f'}{f} = S$$

$$\Rightarrow \boxed{f' = S f}$$

Note
 (+) It may be noted that as the load is applied, the natural effect of the load is to slow down the motor. Hence the slip (diff. of N_s & N_r) increases and with it increases the current & torque, till the driving torque of the motor balances the retarding torque of the load.

\sqrt{f}

$\phi = \omega_s P_s T = T$

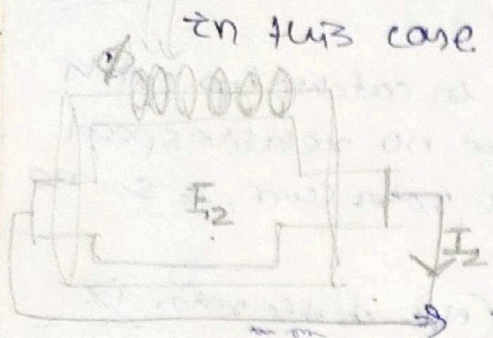
Torque equation →

The torque of motor depends on the interaction of strength of field and frequency of field and frequency of rotor current. Thus the torque developed by the motor is directly proportional to (i) rotor current, (ii) stator flux per pole, and (iii) power factor of rotor circuit.

In case of dc motor

$$T \propto \phi I_2 \cos \phi_2$$

where I_2 → rotor current
 ϕ_2 → angle betⁿ rotor emb^s & rotor current
 since rotor emb/phase @ standstill
 where k → constant. $E \propto \phi$



$$T = k \phi I_2 \cos \phi_2$$

as $E_2 \propto \phi$

where E_2 → rotor emb per phase at standstill

$$T \propto E_2 I_2 \cos \phi_2$$

$$T = k_1 E_2 I_2 \cos \phi_2$$

Starting torque (T_{st}) →

Let E_2 → rotor emb/phase at standstill.

X_2 → rotor reactance/phase at "

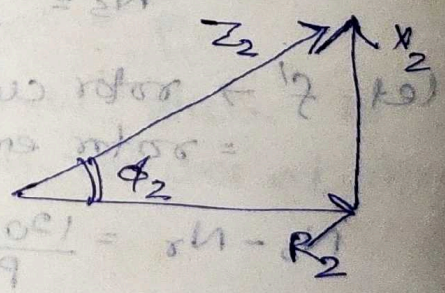
R_2 → rotor resistance/phase

Z_2 → rotor impd/phase at standstill.

$$Z_2 = \sqrt{R_2^2 + X_2^2}$$

$$\cos \phi_2 = \frac{R_2}{Z_2}$$

$$= \frac{R_2}{\sqrt{R_2^2 + X_2^2}}$$



I_2 → rotor current/phase at standstill

$$= \frac{E_2}{Z_2} = \frac{E_2}{\sqrt{R_2^2 + X_2^2}}$$

$$T = k_1 E_2 I_2 \cos \phi_2$$

$$= k_1 \cdot E_2 \cdot \frac{E_2}{\sqrt{R_2^2 + X_2^2}} \cdot \frac{R_2}{\sqrt{R_2^2 + X_2^2}}$$

$$T_{st} = k_1 R_2 E_2^2 / (R_2^2 + X_2^2)$$

experimentally. $K_1 = \frac{3}{(2\pi \cdot \frac{N_s}{60})}$

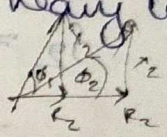
$N_s = \frac{120f}{p}$ rpm.



starting torque of squirrel cage motor.

stator current I₂ very large but st. torque very poor

The resistance of a squirrel cage motor is fixed and small as compared to reactance which is very large. Hence the starting current I_2 though very large in magnitude lags by a very large angle behind E_2 , which with a result that starting torque per ampere is very poor. Hence these motors aren't very useful, unless the motor has to start against heavy loads.



Starting torque of a slip ring motor.

Starting torque of such a motor is increased by improving p.f by adding external resistance from star connected rheostat.

Condition for max^m starting torque (T_{st})

$$T_{st} \propto \frac{E_2^2 R_2}{(R_2^2 + X_2^2)}$$

T_{st} is max^m when $\frac{dT_{st}}{dR_2} = 0$

$$\Rightarrow \frac{E_2^2 [(R_2^2 + X_2^2) \cdot 1 - R_2 \cdot 2R_2]}{(R_2^2 + X_2^2)^2} = 0$$

$$\Rightarrow -R_2^2 + X_2^2 = 2R_2^2$$

$$\Rightarrow X_2^2 = 3R_2^2$$

$$\Rightarrow R_2 = X_2$$

$$T_{st} (\text{max}) = \frac{K_1 E_2^2}{2R_2} = \frac{K_1 E_2^2}{2X_2}$$

R₂ = X₂ for max starting torque

Effect of change in supply volt. on starting torque

$$\text{As } E_2 \propto V \text{ (supply volt.)}$$

$$\text{So } T_{st} \propto \frac{V^2 R_2}{R_2^2 + X_2^2}$$

$$\Rightarrow \boxed{T_{st} \propto V^2}$$

Rotor emf & reactance under running cond.

Let $E_2 \rightarrow$ rotor induced emf/phase at standstill

$X_2 \rightarrow$ rotor reactance/phase at standstill

$f_2 =$ rotor current freq. at standstill = f (supply)

When the rotor is stationary i.e. $s = 1$, the freq. of rotor emf is the same as the supply frequency.

The value of emf induced in the rotor at standstill is max^m becoz the relative speed betⁿ the rotor & the revolving stator flux is max^m.

When rotor starts running the relative speed betⁿ the rotating stator flux is decreased. Hence the rotor induced emf which is directly proportional to this relative speed is also decreased. Hence for a slip 's' the rotor emf will be 's' times the induced emf at standstill i.e.

$E_r =$ rotor emf/phase under running cond.

$$\boxed{E_r = s E_2}$$

$X_r =$ rotor reactance/phase under run

$$\text{Cond.} = \boxed{s X_2}$$

Torque under running cond. \rightarrow

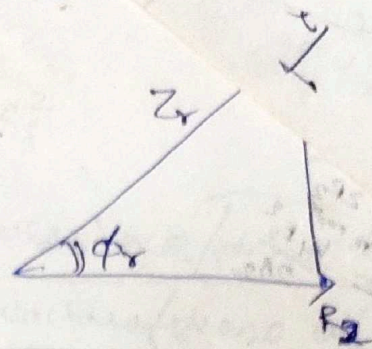
$$T \propto E_r I_r \cos \phi_r$$

$$\frac{s E_2}{s X_2} = (\text{const}) \frac{1}{s}$$

~~$T \propto \phi I$~~
 $T \propto \phi I_r \cos \phi_r$

$$I_r = \frac{E_2}{Z_r}$$

$$\frac{SE_2}{\sqrt{R_2^2 + (SX_2)^2}}$$



$$\cos \phi_r = \frac{R_2}{Z_r} = \frac{R_2}{\sqrt{R_2^2 + (SX_2)^2}}$$

So $T_r \propto \phi SE_2 \frac{R_2}{\sqrt{R_2^2 + (SX_2)^2}}$

$$\Rightarrow T_r \propto \frac{\phi SE_2 R_2}{R_2^2 + (SX_2)^2}$$

As $E_2 \propto \phi$

$$\Rightarrow T_r \propto \frac{SE_2^2 R_2}{R_2^2 + (SX_2)^2}$$

$$\Rightarrow T_r = k_1 \frac{SE_2^2 R_2}{R_2^2 + (SX_2)^2}$$

where $k_1 = \frac{3}{2\pi \frac{N_s}{60}}$

Condition for max^m torque under running condⁿ:

$$T_r = k_1 \frac{SE_2^2 R_2}{R_2^2 + (SX_2)^2}$$

T_r is max^m when $\frac{dT_r}{dS} = 0$

$$\Rightarrow k_1 E_2^2 R_2 \left[R_2^2 + (SX_2)^2 \cdot \frac{d(SX_2)^2}{dS} - S[2SX_2^2] \right] = 0$$

$$\Rightarrow R_2^2 + (SX_2)^2 = 2S^2 X_2^2$$

$$\Rightarrow R_2^2 = (SX_2)^2$$

$$\Rightarrow R_2 = SX_2$$

Effect of \max^m

$$= K_1 \frac{SE_2^2}{2R_2^2}$$

$$= K_1 \frac{SE_2^2}{2R_2}$$

$$= K_1 \frac{SE_2^2}{2sX_2}$$

start $T_{start} = \frac{K_1 E_2^2}{2R_2}$

run $T_{run} = \frac{K_1 E_2^2}{2R_2}$

$T_{start} \neq T_{run}$

$T_{start} = \frac{K_1 E_2^2}{2R_2}$

$T_{run} = \frac{K_1 E_2^2}{2sX_2}$

$$T_r(\max^m) = \frac{K_1 E_2^2}{2X_2}$$

- (i) \max^m torque is independent of rotor resistance
- (ii) But the speed or slip at which \max^m torque occurs is determined by the rotor resistance.
- (iii) By varying rotor resistance (possible only with slip ring motor) \max^m torque can be made to occur at any desired slip or motor speed.
- (iv) \max^m torque varies inversely with standstill reactance, hence it should be kept as small as possible.
- (v) \max^m torque varies directly as the sq. of applied voltage.
- (vi) For obtaining \max^m torque at starting, rotor resistance must be equal to rotor reactance (by putting $s=1$)

F/L torque & \max^m torque: \rightarrow

Let $S_f \rightarrow$ slip corresponding to F/L torque

$$T_f \propto \frac{S_f R_2}{R_2^2 + (S_f X_2)^2}$$

$$T_{max} \propto \frac{1}{2X_2}$$

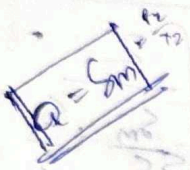
$$\Rightarrow \frac{T_f}{T_{max}} = \frac{2S_f R_2 X_2}{R_2^2 + (S_f X_2)^2}$$

$$\Rightarrow \frac{T_s}{T_{max}} = \frac{2S_f \frac{R_2}{X_2}}{\left(\frac{R_2}{X_2}\right)^2 + S_f^2}$$

Let $a = \frac{R_2}{X_2}$ = $\frac{\text{Rotor resistance/phase at standstill}}{\text{Rotor reactance/phase at standstill}}$

$$\Rightarrow \frac{T_s}{T_{max}} = \frac{2aS_f}{a^2 + S_f^2}$$

Also $\frac{R_2}{X_2} = S_m = \text{slip corresponding to max}^m \text{ torque}$



$$\left(\frac{T_s}{T_{max}} = \frac{2S_m S_f}{S_m^2 + S_f^2} \right) = \frac{2}{\frac{S_m}{S_f} + \frac{S_f}{S_m}} = a$$

Starting torque & max^m torque: \rightarrow

$$T_{st} \propto \frac{R_2}{R_2^2 + X_2^2}$$

$$T_{max} \propto \frac{1}{2X_2}$$

$$\frac{T_{st}}{T_{max}} = \frac{2R_2 X_2}{R_2^2 + X_2^2} = \frac{2R_2/X_2}{\left(\frac{R_2}{X_2}\right)^2 + 1}$$

take $\frac{R_2}{X_2} = a = S_m$

$$\Rightarrow \frac{T_{st}}{T_{max}} = \frac{2a}{a^2 + 1} = \frac{2S_m}{S_m^2 + 1} \quad \text{--- (2)}$$

$$\frac{T_s}{T_{max}} \times \frac{T_{max}}{T_{st}} = \frac{2aS_f}{a^2 + S_f^2} \cdot \frac{a^2 + 1}{2a}$$

$$\left(\frac{T_s}{T_{st}} \right) = \frac{(a^2 + 1)S_f}{a^2 + S_f^2} \quad \text{--- (3)}$$

$$\frac{T_f}{T_{max}} = \frac{2a s_f}{a^2 + s_f^2} = \frac{2}{\left(\frac{a^2 + s_f^2}{a s_f}\right)}$$

$$= \frac{2}{\left(\frac{a^2}{a s_f} + \frac{s_f^2}{a s_f}\right)}$$

$$= \frac{2}{\left(\frac{a}{s_f} + \frac{s_f}{a}\right)}$$

$$T_f = T_{max} \left[\frac{2}{\frac{s_m}{s_f} + \frac{s_f}{s_m}} \right] = T_b \times \left[\frac{2}{\frac{s_m}{s_f} + \frac{s_f}{s_m}} \right]$$

where T_b = Breakdown torque

= T_{max} / Pull out torque.

$$** S_b = s_m = a = \frac{R_2}{X_2}$$

Effect of change in supply volt. on Torque & speed

$$\text{Torque} \propto \frac{SE_2^2 R_2}{R_2^2 + (sX_2)^2}$$

$$\text{as } E_2 \propto V$$

$$T \propto sV^2$$

consequently breq. $T \propto \frac{sV^2}{f}$

$$\frac{s s}{1+s} = \frac{t s}{x s}$$

$$\frac{1+s}{s} \cdot \frac{2s}{2+s} = \frac{x s}{t s} \times \frac{t}{x}$$